

Lecture 20

- How to tune a PID based on empirical methods? -

Review

- PID (proportional–integral–derivative controller)
- Basic properties
- PI implementation
- PD implementation
- PID implementation

Today

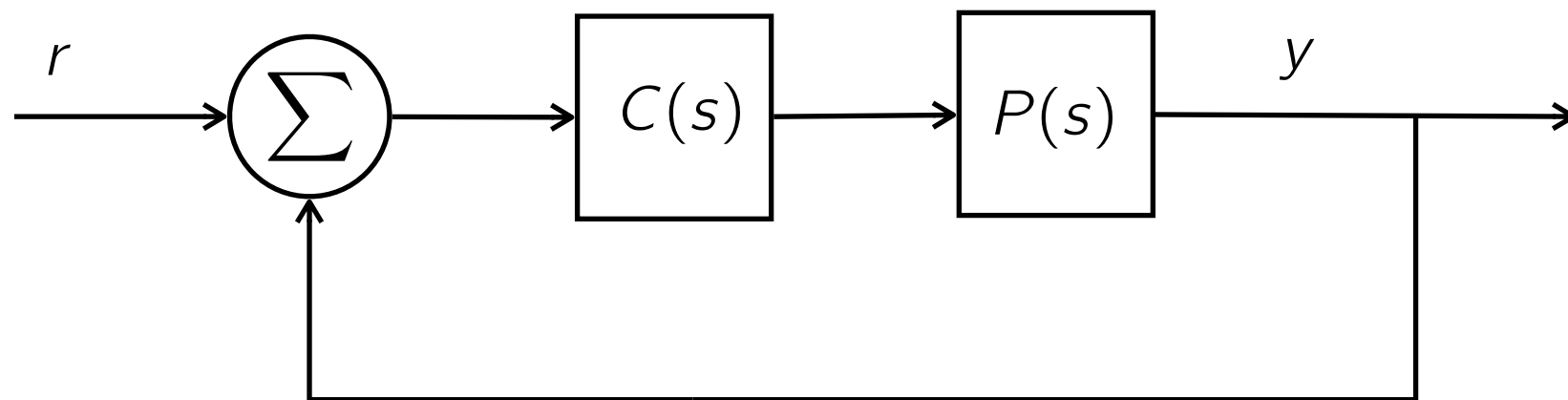
- PID controller (3rd order)
- Empirical tuning methods

Problem

Consider a system with the following system:

$$P(s) = \frac{1}{s^2 + 2s}$$

$$C(s) = \frac{k_d s^2 + k_p s + k_i}{s}$$



Design a PID controller such that:

- Overshoot percentage of 20%
- $T_s = 1$ second

PID controller

- Compute closed-loop transfer function

$$\begin{aligned}\frac{C(s)P(s)}{1 + C(s)P(s)} &= \frac{\frac{k_d s^2 + k_p s + k_i}{s} \frac{1}{s^2 + 2s}}{1 + \frac{k_d s^2 + k_p s + k_i}{s} \frac{1}{s^2 + 2s}} \\ &= \frac{k_d s^2 + k_p s + k_i}{s(s^2 + 2s) + (k_d s^2 + k_p s + k_i)}\end{aligned}$$

- Steady state error (closed-loop)

$$\begin{aligned}e_{ss} &= \lim_{s \rightarrow 0} \left[1 - \frac{C(s)P(s)}{1 + C(s)P(s)} \right] \\ &= \lim_{s \rightarrow 0} \left[1 - \frac{k_d^2 + k_p s + k_i}{s(s^2 + 2s) + (k_d s^2 + k_p s + k_i)} \right] \\ &= 1 - \frac{k_i}{k_i} = 0\end{aligned}$$

- Third order characteristic equation

$$\begin{aligned}\Delta &= s(s^2 + 2s) + (k_d s^2 + k_p s + k_i) \\ &= s^3 + (2 + k_d)s^2 + k_p s + k_i\end{aligned}$$

PID controller

- Consider a second order system
- Add a third pole
- Third pole further away from the imaginary axis
- Pole location for second order system

$$\begin{aligned} p_{1,2} &= -\frac{2\zeta\omega_0 \pm \sqrt{4\zeta^2\omega_0^2 - 4\omega_0^2}}{2} \\ &= -\zeta\omega_0 \pm \omega_0\sqrt{\zeta^2 - 1} \end{aligned}$$

- Third pole at

$$p_3 < -\zeta\omega_0$$

Meet system specifications based on behavior of second order system

PID controller

- Desired behavior of second order system

$$\frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2}$$

Settling time

$$T_s = 1 \approx \frac{4}{\zeta\omega_0} \implies \zeta\omega_0 = 4$$

Percentage of overshoot

$$\begin{aligned} P.O. = 0.2 &= 100e^{-\frac{\zeta\pi}{\sqrt{1-\zeta^2}}} \\ &\implies \zeta = 0.89 \\ &\implies \omega_0 = 4.5 \end{aligned}$$

- Adding a third pole at $p_3 = -10$

$$\begin{aligned} \frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2} \frac{1}{s - p_3} &= \frac{20.25}{s^2 + 8s + 20.25} \frac{1}{s + 10} \\ &= \frac{20.25}{s^3 + 18s^2 + 100.25s + 202.5} \end{aligned}$$

← desired characteristic equation

PID controller

$$\Delta_{\text{desired}} = \Delta$$

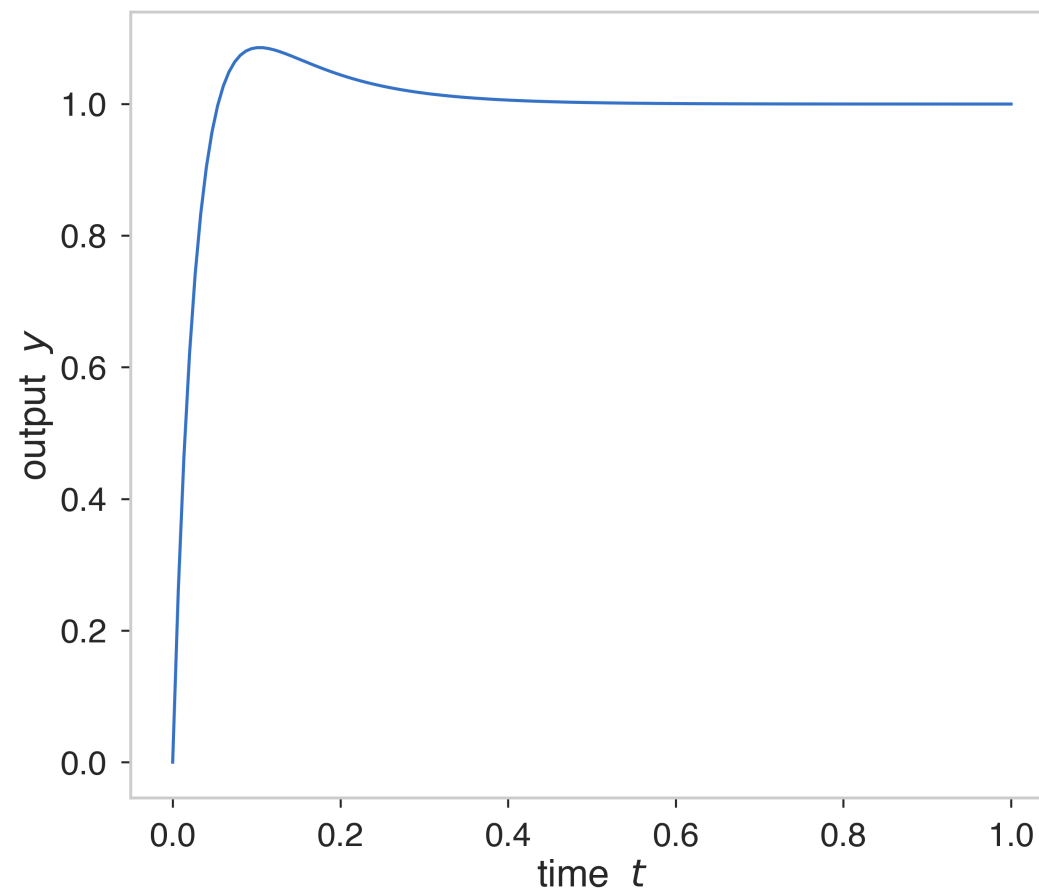
$$s^3 + 18s^2 + 100.25s + 202.5 = s^3 + (2 + k_d)s^2 + k_p s + k_i$$

Controller gains

$$k_p = 100.25$$

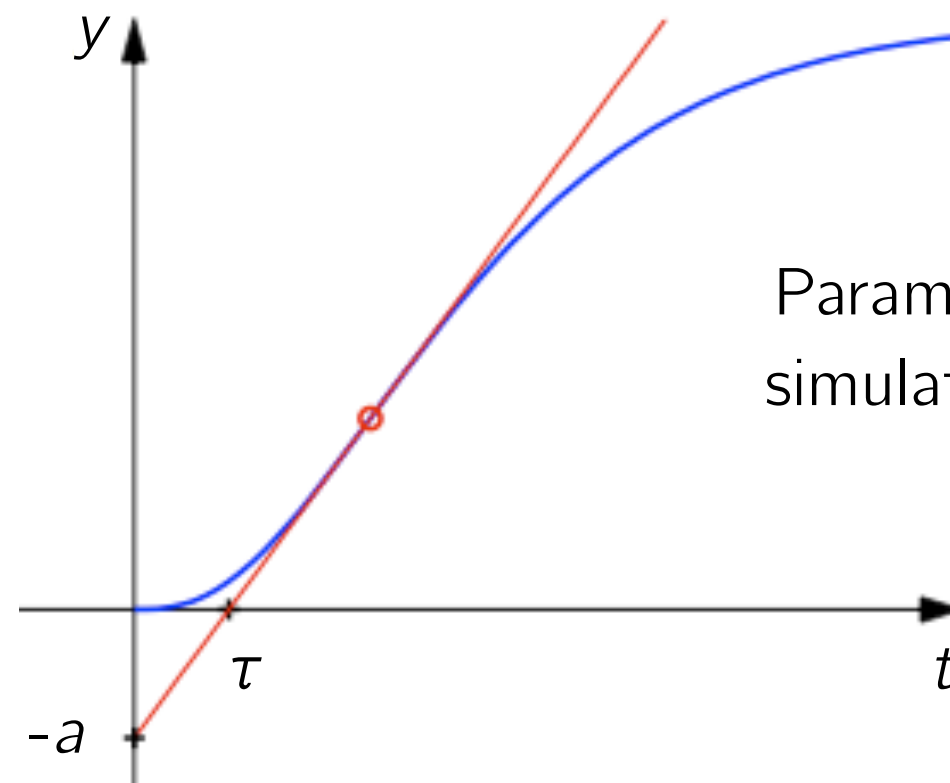
$$k_d = 18 - 2 = 16$$

$$k_i = 202.5$$



The time domain method

- Open loop unit step response of the process
- Steepest slope of the step response: a/τ
- The parameter τ is an approximation of the time delay of the system



Parameters obtained by extensive simulation of a range of processes

Type	k_p	T_i	T_d
P	$1/a$		
PI	$0.9/a$	3τ	
PID	$1.2/a$	2τ	0.5τ

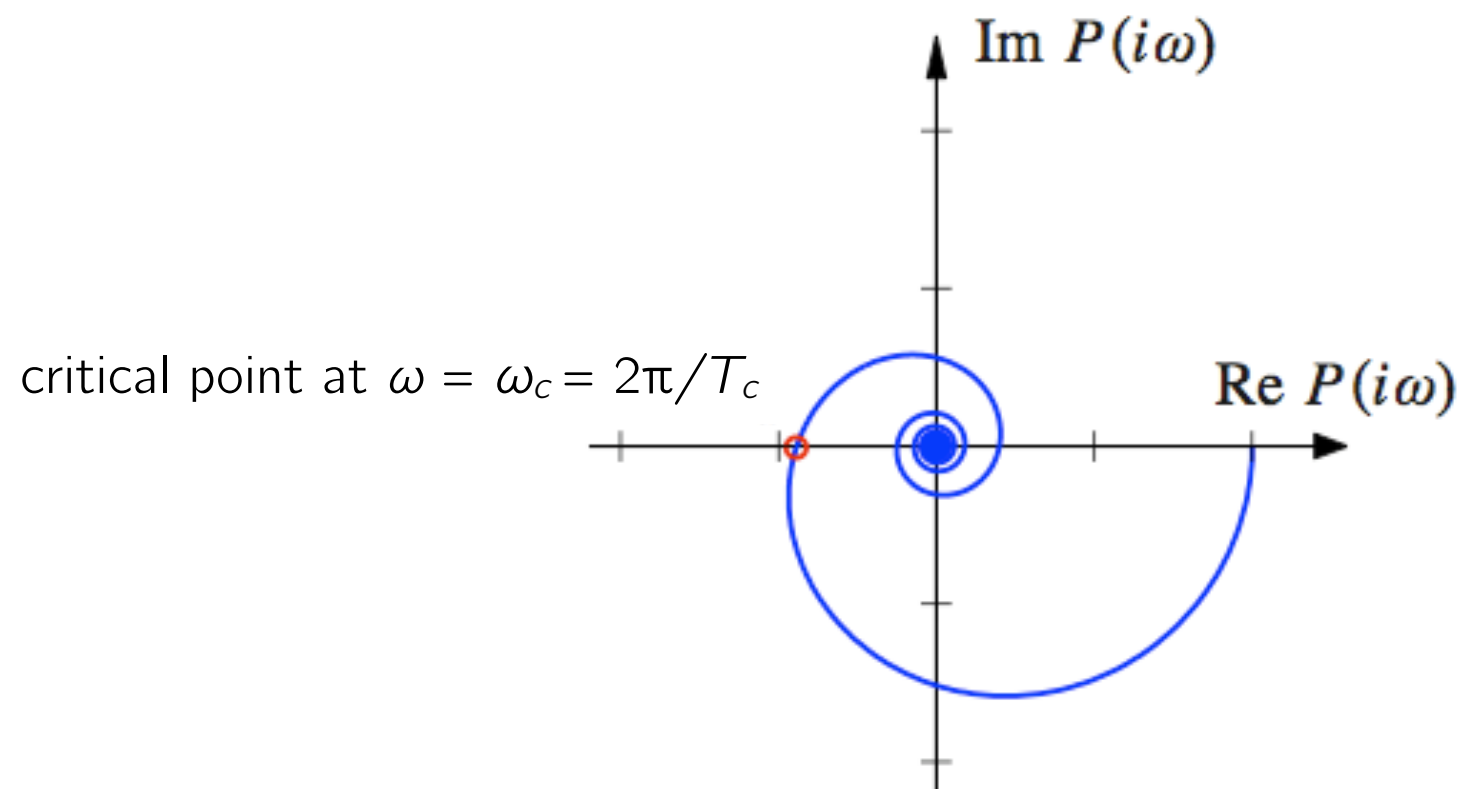
and

$$k_i = \frac{k_p}{T_i}$$

$$k_d = k_p T_d$$

Frequency domain method

- Controller's integral and derivative gains are set to zero
- Proportional gain is increased until the system starts to oscillate
- Critical value of $k_p = k_c$ is observed together with the period of oscillation T_c



Type	k_p	T_i	T_d
P	$0.5k_c$		
PI	$0.4k_c$	$0.8T_c$	
PID	$0.6k_c$	$0.5T_c$	$0.125T_c$

Ziegler–Nichols method